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ON HIGHLY OSCILLATORY PROBLEMS ARISING IN ELECTRONIC ENGINEERING

Marissa Condon¹, Alfredo Deaño² and Arieh Iserles²

Abstract. In this paper, we consider linear ordinary differential equations originating in electronic engineering, which exhibit exceedingly rapid oscillation. Moreover, the oscillation model is completely different from the familiar framework of asymptotic analysis of highly oscillatory integrals. Using a Bessel-function identity, we expand the oscillator into asymptotic series, and this allows us to extend Filon-type approach to this setting. The outcome is a time-stepping method that guarantees high accuracy regardless of the rate of oscillation.

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1. Introduction

The focus of our attention in this paper is the discretization of ordinary differential equations of the form

$$\mathbf{y}' = A\mathbf{y} + E(t)\mathbf{g}(t), \quad t \ge 0, \qquad \mathbf{y}(0) = \mathbf{y}_0 \in \mathbb{R}^d,$$
 (1.1)

where A is a $d \times d$ matrix, \mathbf{g} is a d-vector of functions while E is a $d \times d$ matrix function, $E_{k,l}(t) = \chi_{k,l} e^{\tau_{k,l} \sin \omega_{k,l} t}$, $k, l = 1, \ldots, d$. While we may assume that the eigenvalues of A are of moderate size, the terms of E are highly oscillatory, since we allow for $\max \omega_{k,l} \gg 1$. Moreover, it is perfectly possible for different frequencies $\omega_{k,l}$ to differ in size by many orders of magnitude.

Equation (1.1) is a much-simplified model of more complicated, in general nonlinear, ordinary differential and differential-algebraic equations originating in electronic engineering. In subsequent papers we plan to address ourselves, using and generalizing techniques introduced in this paper in tandem with other relevant methodologies (exponential integrators, waveform relaxation), to more realistic systems of nonlinear ODEs and DAEs.

High-frequency signals abound in Radio Frequency (RF) communication systems. This is a consequence of the need for modulation: the imposition of a lower-frequency information signal onto a high-frequency carrier. The goal is to enable antennae of a manageable size to be employed for audio transmission. Antennae of the order of several miles to several thousand miles would be required if modulation was not performed. In RF communication systems, signals in the MHz frequency range and higher are common. Furthermore,

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¹ School of Electronic Engineering, Dublin City University, Dublin 9, Ireland.

² Department of Applied Mathematics and Theoretical Physics, Centre for Mathematical Sciences, University of Cambridge, Wilberforce Rd, Cambridge CB3 0WA, UK. A.Iserles@damtp.cam.ac.uk

nonlinearities abound in RF transmission systems owing to the presence of solid-state amplifiers, mixers and so on [13].

Most RF systems involve a linear part and a nonlinear part with the linear part due to the presence of linear resistors, inductors and capacitors and the nonlinear part due to amplifiers, mixers or nonlinear and controlled resistors and capacitors. Equations (1.1) are a simplified model with many of the nonlinearities approximated by linear terms. The occurrence of the $e^{\tau_{k,l}\sin\omega_{k,l}t}$ is due to the input of sine-waves to terminals of circuits with diodes or transistors. Note that sinusoids and amplifiers involving transistors are ubiquitous in RF systems [6,19].

The recent explosion of developments in the RF and telecommunications industry has put pressure on circuit designers for faster simulations, faster designs and faster product output and the existing Computer Aided Design (CAD) tools have struggled to keep pace. In addition, the growing complexity of the modulation formats is rendering software tools unacceptably slow and consequently, unsatisfactory. There is therefore, an urgent need for a complete revamp and update of the fundamental numerical processes within these CAD packages taking into account the modern developments and formats.

Some recent work in this direction is that by e.g. [18] and subsequent work by Pulch [17] and Dautbegovic $et \ al.$ [3]. However, much more work is required to generate algorithms that are well-suited and effective for the application areas in hand.

In this paper we commence from the most basic setting, striving to create a framework for an effective solution of highly oscillatory equations of electronic engineering. We use a simplified setting to develop and analyse a range of tools which will be deployed in future work in more challenging and realistic situation.

On the face of it, solving (1.1) is trivial, because we can write the solution of this linear system explicitly as variation of constants,

$$\mathbf{y}(t_{n+1}) = e^{hA} \mathbf{y}(t_n) + \int_{t_n}^{t_{n+1}} e^{(t_{n+1} - \xi)A} E(\xi) \mathbf{g}(\xi) \,d\xi,$$
(1.2)

where $t_{n+1} = t_n + h$. This, however, is not a very helpful observation because of the presence of highly oscillatory terms inside the integral. Specifically, rewriting (1.2) component-wise, we have

$$y_k(t_{n+1}) = \sum_{i=1}^d F_{k,i}(h)y_i(t_n) + \sum_{i=1}^d \sum_{j=1}^d \chi_{i,j} \int_{t_n}^{t_{n+1}} F_{k,i}(t_{n+1} - \xi) e^{\tau_{i,j} \sin \omega_{i,j} \xi} g_j(\xi) d\xi$$

for k = 1, ..., d, where $F(t) = e^{tA}$. While the computation of the matrix exponential is standard, the intrinsic difficulty is represented by practical computation of integrals of the form

$$\int_{t_n}^{t_{n+1}} F_{k,i}(t_{n+1} - \xi) e^{\tau_{i,j} \sin \omega_{i,j} \xi} g_j(\xi) d\xi = \frac{1}{2} h \int_{-1}^1 F_{k,i}(\frac{1}{2}h(1-x)) g_j(t_n + \frac{1}{2}h(1+x)) e^{\tau_{i,j} \sin \omega_{i,j} \left(t_n + \frac{1}{2}h(1+x)\right)} dx$$
(1.3)

for $\omega_{i,j} \gg 1$. Since classical numerical methods for non-oscillatory integrals, e.g. Gaussian quadrature, require the decomposition of the integration interval into $\mathcal{O}(\omega)$ sub-panels [4], and recalling that we have d^3 such intervals in each step, they are completely unfit for purpose.

An alternative is provided by contemporary methods for highly oscillatory quadrature, an area that has undergone significant developments in the last few years. The problem, though, is that the integral (1.3) does not fit into the framework of traditional asymptotic theory for highly oscillatory integrals [7,20]: the latter is concerned with integrals of the form $\int_{\Omega} f(\boldsymbol{x}) \mathrm{e}^{\mathrm{i}\omega g(\boldsymbol{x})} \,\mathrm{d}\boldsymbol{x}$, where $\omega \gg 1$ while neither f nor g are oscillatory. This is also the case with the methods for numerical calculation of highly oscillatory integrals that have been developed recently [6,10,16].

Yet another approach is to disregard the explicit formula (1.2) and use exponential integrators to solve the system (1.1). This is not very promising either. Most exponential integrators designed to cope with high oscillations do this in a Hamiltonian setting, which does not fit the paradigm of (1.1) [5]. Moreover, they are not designed to deal with the multiscale nature of (1.1) and with truly huge frequencies $\omega_{i,j}$ therein. An exception

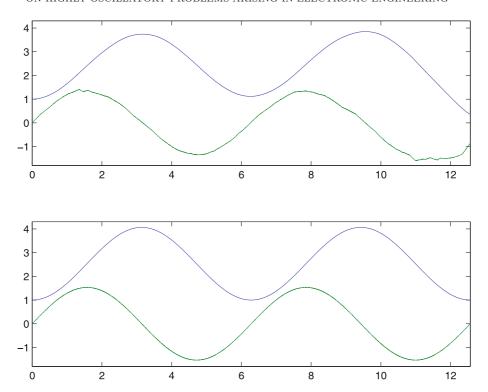


FIGURE 1. The numerical (top, with $RelTol = 10^{-4}$) and true (bottom) solution of (1.4) in the first two periods.

TABLE 1. The performance of ode45 in the interval $[0, 4\pi]$ for different relative error tolerances for the system (1.4) with $\omega = 10^4$.

| RelTol | Number of steps | Numerical error | | |
|-----------|-----------------|-----------------|---------------|--|
| | | in $y(4\pi)$ | in $y'(4\pi)$ | |
| 10^{-4} | 61 441 | -6.42_{-01} | -8.57_{-01} | |
| 10^{-5} | 123405 | 9.61_{-04} | -2.19_{-02} | |
| 10^{-6} | 240645 | -1.01_{-04} | 4.57_{-04} | |
| 10^{-7} | 377057 | 1.94_{-06} | 5.17_{-05} | |

to the Hamiltonian setting is provided in [14], but this does not advance us much since it takes us to the very same highly oscillatory quadrature methods which we have already deemed unsuitable in the last paragraph.

Finally, we can disregard the special structure of (1.1) and just use an all-purpose ODE solver, placing our trust in its error-control and variable-step strategies. Thus, we have solved the system

$$y'' + y = 2e^{\sin \omega t}, \quad t \ge 0, \qquad y(0) = 1, \quad y'(0) = 0,$$
 (1.4)

with the Matlab routine ode45, employing different error tolerances and setting $\omega=10\,000$. The solution of (1.4) is periodic with period 2π and we have examined a numerical solution across two periods. We have set different values of the relative error tolerance parameter RelTol, setting in each case AbsTol = 10^{-3} × RelTol.

In Figure 1 we present a numerical solution (admittedly, with the least relative error, RelTol = 10^{-4} , yet tenfold smaller than the MATLAB default) of (1.4), comparing it with the exact solution. It is evident that the quality of the numerical solution deteriorates fairly rapidly. Cursory examination of the exact solution might

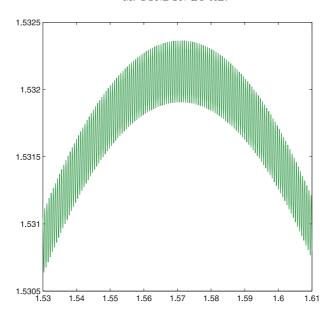


FIGURE 2. A close-up on y' within a narrow window, exhibiting rapid small-amplitude oscillations.

be misleading, since it appears to be a very 'nice' function, varying in a sedate manner. However, once we magnify the solution within a short window, as in Figure 2, we note that it exhibits very rapid, small-amplitude oscillations. Such oscillations are bound to inhibit the step size in any standard error-control mechanism in all-purpose software and this, perhaps unsurprisingly, is reflected in Table 1. Another important observation is that the numerical (absolute) error falls substantially short of either relative or absolute error-tolerance parameters. This breakdown in error control has been already reported for other highly oscillatory ODE systems [9]. Note that (1.4) is a toy problem, not just because we are interested in larger systems with many frequencies, but also because $\omega = 10^4$ is a fairly small frequency within our framework. Realistic electronic circuits are likely to exhibit fast oscillations in the range of $\approx 10^8$. This, clearly, is beyond the scope of any standard ODE software.

The solution that we propose in this paper is to analyse the asymptotic behaviour of the integral (1.3), thereby creating the right tools for the extension of Filon-type quadrature [11] to this setting. This will lead not just to a practical algorithm for the calculation of (1.1) with arbitrarily large frequencies $\omega_{i,j}$ (indeed, the higher the frequency, the better!) but will also serve us in future generalization of this equation to full nonlinear setting. Finally, asymptotic expansion and numerical computation of the highly oscillatory integral (1.3) and, in future publication, of its generalisations is of an independent mathematical interest.

2. The asymptotics of the ExpSin integral

Mindful of (1.3), we are concerned with the asymptotic behaviour of the integral

$$I[f] = \int_{-1}^{1} f(x)e^{\tau \sin \omega(\alpha x + \beta)} dx, \qquad (2.1)$$

where $\alpha, \beta \in \mathbb{R}$, $\tau \in \mathbb{C} \setminus \{0\}$ and $\omega \gg 1$. For a want of a better name, we call (2.1) the *ExpSin integral*. Even the briefest examination of (2.1) highlights a crucial difference between the ExpSin integral and the 'standard model' of asymptotic theory of highly oscillatory integrals. Thus, suppose that we move the ω to front

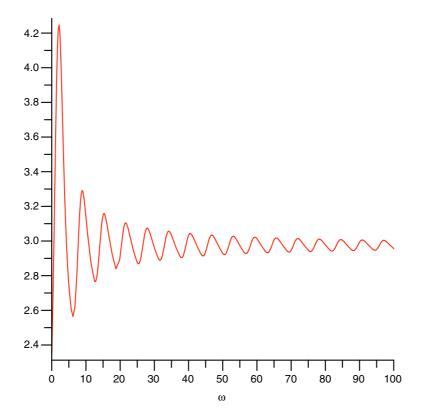


FIGURE 3. The integral $I[e^x]$ for $\alpha = 1$, $\beta = 0$ and $0 \le \omega \le 100$.

of the sine function. It follows at once from the method of stationary phase [20] that

$$\int_{-1}^{1} f(x) e^{\tau \omega \sin(\alpha x + \beta)} dx = \mathcal{O}\left(\omega^{-\frac{1}{2}}\right), \qquad \omega \gg 1,$$

provided that $[-\beta + (m + \frac{1}{2})\pi]/\alpha \in [-1, 1]$ for some $m \in \mathbb{Z}$,

$$\int_{-1}^{1} f(x) e^{\tau \omega \sin(\alpha x + \beta)} dx = \mathcal{O}(\omega^{-1}), \qquad \omega \gg 1,$$

otherwise. On the other hand, for $\tau \in \mathbb{R}$ and f(x) > 0, $x \in [-1, 1]$, it follows at once that

$$0 < 2\mathrm{e}^{-1} \, \min_{x \in [-1,1]} f(x) \le \boldsymbol{I}[f] \le 2\mathrm{e} \, \max_{x \in [-1,1]} f(x)$$

and the integral is bounded away from zero uniformly in $\omega \in \mathbb{R}$. This is demonstrated in Figure 3. The key step toward the analysis of the ExpSin integral is the identity

$$e^{\tau \sin \theta} = I_0(\tau) + 2\sum_{k=0}^{\infty} (-1)^k I_{2k+1}(\tau) \sin(2k+1)\theta + 2\sum_{k=1}^{\infty} (-1)^k I_{2k}(\tau) \cos 2k\theta, \tag{2.2}$$

where I_k is the kth modified Bessel function [1], p. 376, formula (9.6.35). Letting $\theta = \omega(\alpha x + \beta)$ in (2.1), we thus obtain

$$I[f] = I_0(\tau) \int_{-1}^1 f(x) dx + 2 \sum_{k=0}^{\infty} (-1)^k I_{2k+1}(\tau) \int_{-1}^1 f(x) \sin((2k+1)\omega(\alpha x + \beta)) dx$$
$$+ 2 \sum_{k=1}^{\infty} (-1)^k I_{2k}(\tau) \int_{-1}^1 f(x) \cos(2k\omega(\alpha x + \beta)) dx.$$
(2.3)

We thus express I[f] as an infinite sum of integrals, all of which (except for the first) are themselves highly oscillatory. Before we expand these integrals in turn, it is useful to comment further about this sum. Since all oscillatory integrals are o(1) for $\omega \gg 1$, we deduce that

$$\lim_{\omega \to \infty} \mathbf{I}[f] = I_0(\tau) \int_{-1}^1 f(x) \, \mathrm{d}x.$$

Moreover, we can deduce at once from [1], p. 365, formula (9.3.1), that

$$I_k(\tau) \sim \frac{1}{\sqrt{2\pi k}} \left(\frac{\mathrm{e}\tau}{2k}\right)^k, \qquad k \gg 1.$$

Since the highly oscillatory integrals are small (as we will see soon, they are $\mathcal{O}(\omega^{-1})$), we conclude that the infinite series converge very rapidly, at a spectral speed.

Let

$$C_{\sigma,\rho}[f] = \int_{-1}^{1} f(x)\cos(\sigma x + \rho) dx, \qquad S_{\sigma,\rho}[f] = \int_{-1}^{1} f(x)\sin(\sigma x + \rho) dx,$$

therefore (2.3) becomes

$$I[f] = I_0(\tau) \int_{-1}^{1} f(x) dx + 2 \sum_{k=0}^{\infty} (-1)^k I_{2k+1}(\tau) S_{(2k+1)\omega\alpha,(2k+1)\omega\beta}[f]$$

$$+ 2 \sum_{k=1}^{\infty} (-1)^k I_{2k}(\tau) C_{2k\omega\alpha,2k\omega\beta}[f].$$
(2.4)

Let us assume that $f \in C^{\infty}[-1,1]$. It is fairly straightforward, although laborious, to expand $C_{\sigma,\rho}[f]$ and $S_{\sigma,\rho}[f]$ asymptotically in inverse powers of $\sigma \neq 0$. The obvious route, letting $C_{\sigma,\rho}[f] + iS_{\sigma,\rho}[f] = \int_{-1}^{1} f(x)e^{i(\sigma x + \rho)} dx$ and using an explicit expansion from [11], is probably less transparent than direct expansion.

Integrating $S_{\sigma,\rho}$, $\sigma \neq 0$, twice by parts we obtain

$$S_{\sigma,\rho}[f] = -\frac{1}{\sigma} \int_{-1}^{1} f(x) \frac{d}{dx} \cos(\sigma x + \rho) dx$$

$$= -\frac{1}{\sigma} [f(1) \cos(\sigma + \rho) - f(-1) \cos(\sigma - \rho)] + \frac{1}{\sigma} \int_{-1}^{1} f'(x) \cos(\sigma x + \rho) dx$$

$$= -\frac{1}{\sigma} [f(1) \cos(\sigma + \rho) - f(-1) \cos(\sigma - \rho)] + \frac{1}{\sigma^2} \int_{-1}^{1} f'(x) \frac{d}{dx} \sin(\sigma x + \rho) dx$$

$$= -\frac{1}{\sigma} [f(1) \cos(\sigma + \rho) - f(-1) \cos(\sigma - \rho)]$$

$$+ \frac{1}{\sigma^2} [f'(1) \sin(\sigma + \rho) + f'(-1) \sin(\sigma - \rho)] - \frac{1}{\sigma^2} S_{\sigma,\rho}[f''].$$

Iterating this expression yields the asymptotic expansion of $S_{\sigma,\rho}[f]$ in inverse powers of σ ,

$$S_{\sigma,\rho}[f] \sim -\sum_{m=0}^{\infty} \frac{(-1)^m}{\sigma^{2m+1}} [f^{(2m)}(1)\cos(\sigma+\rho) - f^{(2m)}(-1)\cos(\sigma-\rho)]$$

$$+\sum_{m=0}^{\infty} \frac{(-1)^m}{\sigma^{2m+2}} [f^{(2m+1)}(1)\sin(\sigma+\rho) + f^{(2m+1)}(-1)\sin(\sigma-\rho)], \qquad \sigma \gg 1.$$
(2.5)

Likewise, using (2.5), we have

$$C_{\sigma,\rho}[f] = \frac{1}{\sigma} \int_{-1}^{1} f(x) \frac{\mathrm{d}}{\mathrm{d}x} \sin(\sigma x + \rho) \, \mathrm{d}x$$

$$= \frac{1}{\sigma} [f(1)\sin(\sigma + \rho) + f(-1)\sin(\sigma - \rho)] - \frac{1}{\sigma} S_{\sigma,\rho}[f']$$

$$\sim \sum_{m=0}^{\infty} \frac{(-1)^m}{\sigma^{2m+1}} [f^{(2m)}(1)\sin(\sigma + \rho) + f^{(2m)}(-1)\sin(\sigma - \rho)]$$

$$+ \sum_{m=0}^{\infty} \frac{(-1)^m}{\sigma^{2m+2}} [f^{(2m+1)}(1)\cos(\sigma + \rho) - f^{(2m+1)}(-1)\cos(\sigma - \rho)], \qquad \sigma \gg 1.$$
(2.6)

Substituting (2.5) and (2.6) into (2.4) results in

$$\begin{split} I[f] &\sim I_0(\tau) \int_{-1}^1 f(x) \, \mathrm{d}x \\ &+ 2 \sum_{k=0}^\infty (-1)^k I_{2k+1}(\tau) \left\{ - \sum_{m=0}^\infty \frac{(-1)^m}{[(2k+1)\omega\alpha]^{2m+1}} [f^{(2m)}(1) \cos((2k+1)\omega(\alpha+\beta)) \right. \\ &- f^{(2m)}(-1) \cos((2k+1)\omega(\alpha-\beta))] \right. \\ &+ \sum_{m=0}^\infty \frac{(-1)^m}{[(2k+1)\omega\alpha]^{2m+2}} [f^{(2m+1)}(1) \sin((2k+1)\omega(\alpha+\beta)) \\ &+ f^{(2m+2)}(-1) \sin((2k+1)\omega(\alpha-\beta))] \right\} \\ &+ 2 \sum_{k=1}^\infty (-1)^k I_{2k}(\tau) \left\{ \sum_{m=0}^\infty \frac{(-1)^m}{(2k\omega\alpha)^{2m+1}} [f^{(2m)}(1) \sin(2k\omega(\alpha+\beta)) \right. \\ &+ f^{(2m)}(-1) \sin(2k\omega(\alpha-\beta))] \\ &+ \sum_{m=0}^\infty \frac{(-1)^m}{(2k\omega\alpha)^{2m+2}} [f^{(2m+1)}(1) \cos(2k\omega(\alpha+\beta)) \\ &- f^{(2m+1)}(-1) \cos(2k\omega(\alpha-\beta))] \right\} \\ &= I_0(\tau) \int_{-1}^1 f(x) \, \mathrm{d}x \\ &+ 2 \sum_{m=0}^\infty \frac{(-1)^m}{(\alpha\omega)^{2m+1}} \left[- f^{(2m)}(1) \sum_{k=0}^\infty \frac{(-1)^k I_{2k+1}(\tau)}{(2k+1)^{2m+1}} \cos((2k+1)\omega(\alpha+\beta)) \right. \\ &+ f^{(2m)}(1) \sum_{k=1}^\infty \frac{(-1)^k I_{2k}(\tau)}{(2k)^{2m+1}} \sin(2k\omega(\alpha+\beta)) \\ &+ f^{(2m)}(-1) \sum_{k=0}^\infty \frac{(-1)^k I_{2k+1}(\tau)}{(2k)^{2m+1}} \cos((2k+1)\omega(\alpha-\beta)) \\ &+ f^{(2m)}(-1) \sum_{k=1}^\infty \frac{(-1)^k I_{2k}(\tau)}{(2k)^{2m+1}} \sin(2k\omega(\alpha-\beta)) \right] \\ &+ 2 \sum_{m=0}^\infty \frac{(-1)^m}{(\alpha\omega)^{2m+2}} \left[f^{(2m+1)}(1) \sum_{k=0}^\infty \frac{(-1)^k I_{2k+1}(\tau)}{(2k+1)^{2m+2}} \sin((2k+1)\omega(\alpha+\beta)) \right. \\ &+ f^{(2m+1)}(1) \sum_{k=1}^\infty \frac{(-1)^k I_{2k}(\tau)}{(2k)^{2m+2}} \cos(2k\omega(\alpha+\beta)) \\ &+ f^{(2m+1)}(-1) \sum_{k=0}^\infty \frac{(-1)^k I_{2k+1}(\tau)}{(2k)^{2m+2}} \sin((2k+1)\omega(\alpha-\beta)) \\ &- f^{(2m+1)}(-1) \sum_{k=1}^\infty \frac{(-1)^k I_{2k+1}(\tau)}{(2k)^{2m+2}} \cos(2k\omega(\alpha-\beta)) \right]. \end{split}$$

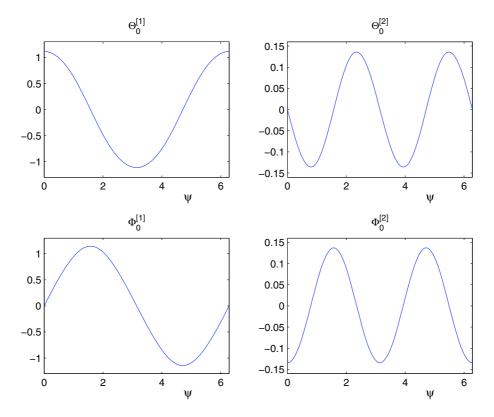


FIGURE 4. The functions $\Theta_0^{[i]}(\psi,1)$ and $\Phi_0^{[i]}(\psi,1)$ for i=1,2 and $0 \le \psi \le 2\pi$.

Let

$$\Theta_m^{[1]}(\psi,\tau) = 2\sum_{k=0}^{\infty} \frac{(-1)^k I_{2k+1}(\tau)}{(2k+1)^{2m+1}} \cos((2k+1)\psi)
\Theta_m^{[2]}(\psi,\tau) = 2\sum_{k=1}^{\infty} \frac{(-1)^k I_{2k}(\tau)}{(2k)^{2m+1}} \sin(2k\psi),
\Phi_m^{[1]}(\psi,\tau) = 2\sum_{k=0}^{\infty} \frac{(-1)^k I_{2k+1}(\tau)}{(2k+1)^{2m+2}} \sin((2k+1)\psi),
\Phi_m^{[2]}(\psi,\tau) = 2\sum_{k=1}^{\infty} \frac{(-1)^k I_{2k}(\tau)}{(2k)^{2m+2}} \cos(2k\psi).$$

Note that the four functions are analytic in ψ, τ for all $m \in \mathbb{Z}_+$ and their convergence is assured. They are periodic in ψ of period 2π for $\Theta^{[1]}$ and $\Phi^{[1]}$, of period π otherwise.

In Figure 4 we display the functions $\Theta^{[i]}_0$ and $\Phi^{[i]}_0$ for i=1,2. Note that the differences between $\Theta^{[i]}_m$ for $m \geq 1$ and $\Theta^{[i]}_0$ (likewise, between $\Phi^{[i]}_m$ and $\Phi^{[i]}_0$) are very small, thus this figure is typical of all ms.

The four functions are infinite series. Yet, the speed of their convergence is so rapid that it is enough to restrict the range of summation to $k \leq 6$ to attain machine accuracy.

Using $\Theta_m^{[i]}$ and $\Phi_m^{[i]}$ we can write conveniently the asymptotic expansion of the ExpSin integral I[f].

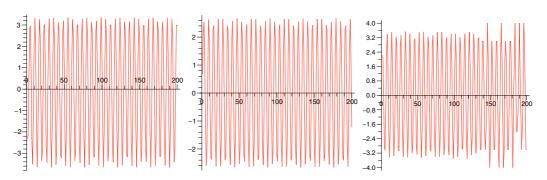


FIGURE 5. Scaled error $\omega^{s+1}|\mathbf{A}_s[\mathrm{e}^x]-\mathbf{I}[\mathrm{e}^x]|$ (with $\alpha=1,\ \beta=0$) for s=1,2,3 for $\omega\in[0,200]$.

Lemma 2.1. Let $\alpha\omega \gg 1$. Then

$$I[f] \sim I_{0}(\tau) \int_{-1}^{1} f(x) dx$$

$$+ \sum_{m=0}^{\infty} \frac{(-1)^{m}}{(\alpha \omega)^{2m+1}} \left\{ f^{(2m)}(1) [\Theta_{m}^{[2]}(\omega(\alpha+\beta)) - \Theta_{m}^{[1]}(\omega(\alpha+\beta))] \right\}$$

$$+ f^{(2m)}(-1) [\Theta_{m}^{[2]}(\omega(\alpha-\beta)) + \Theta_{m}^{[1]}(\omega(\alpha-\beta))] \right\}$$

$$+ \sum_{m=0}^{\infty} \frac{(-1)^{m}}{(\alpha \omega)^{2m+2}} \left\{ f^{(2m+1)}(1) [\Phi_{m}^{[2]}(\omega(\alpha+\beta)) + \Phi_{m}^{[1]}(\omega(\alpha+\beta))] - f^{(2m+1)}(-1) [\Phi_{m}^{[2]}(\omega(\alpha-\beta)) - \Phi_{m}^{[1]}(\omega(\alpha-\beta))] \right\}.$$

$$(2.7)$$

An immediate application of the expansion (2.7) is to the numerical calculation of I[f]. Truncating the series results for $s \in \mathbb{N}$ in the asymptotic method

$$I[f] \approx A_{s}[f] = I_{0}(\tau) \int_{-1}^{1} f(x) dx$$

$$+ \sum_{m=0}^{\lfloor (s-1)/2 \rfloor} \frac{(-1)^{m}}{(\alpha \omega)^{2m+1}} \left\{ f^{(2m)}(1) \left[\Theta_{m}^{[2]}(\omega(\alpha+\beta)) - \Theta_{m}^{[1]}(\omega(\alpha+\beta)) \right] + f^{(2m)}(-1) \left[\Theta_{m}^{[2]}(\omega(\alpha-\beta)) + \Theta_{m}^{[1]}(\omega(\alpha-\beta)) \right] \right\}$$

$$+ \sum_{m=0}^{\lfloor s/2 \rfloor - 1} \frac{(-1)^{m}}{(\alpha \omega)^{2m+2}} \left\{ f^{(2m+1)}(1) \left[\Phi_{m}^{[2]}(\omega(\alpha+\beta)) + \Phi_{m}^{[1]}(\omega(\alpha+\beta)) \right] - f^{(2m+1)}(-1) \left[\Phi_{m}^{[2]}(\omega(\alpha-\beta)) - \Phi_{m}^{[1]}(\omega(\alpha-\beta)) \right] \right\}$$

$$(2.8)$$

and it is trivial to verify that

$$\mathbf{A}_s[f] = \mathbf{I}[f] + \mathcal{O}((\alpha\omega)^{-s-1}), \qquad |\alpha\omega| \gg 1.$$

In a way of an example, we have used (2.8) to compute the integral $I[e^x]$ from Figure 3. The results are displayed in Figure 5 and they confirm the theoretical expectations on asymptotic behaviour. Indeed, they

Table 2. Absolute errors $|\mathbf{A}_s[e^x] - \mathbf{I}[e^x]|$ for s = 1, 2, 3.

| | | | $\omega = 100$ | |
|---|--------------|--------------|---|--------------|
| 1 | 2.14_{-02} | 3.96_{-04} | 1.81_{-04} | 7.39_{-05} |
| 2 | 1.92_{-03} | 2.02_{-05} | 2.22_{-06} | 1.53_{-07} |
| 3 | 2.11_{-04} | 1.44_{-07} | $ \begin{array}{c} 1.81_{-04} \\ 2.22_{-06} \\ 1.76_{-08} \end{array} $ | 1.89_{-09} |

exceed it, because apparently the asymptotic regime sets already for very small $|\alpha\omega|$, rather then only for large frequencies. However, the figure does not exhibit transparently the actual absolute error and for this we refer to Table 2. It is clear that, while for large $|\alpha\omega|$ we need relatively modest values of s, moderate frequencies call for large s and the method becomes expensive: this is only to be expected, because of the asymptotic nature of the method (2.8).

There are two obvious problems associated with the asymptotic method (2.8). Firstly, we said nothing about the numerical evaluation of the leading integral. This, however, can be accomplished easily using classical quadrature, since it is non-oscillatory. Secondly, as demonstrated in Table 2, the formula is useful only for sufficiently large value of $|\alpha\omega|$. Although, in our experience, asymptotic behaviour sets surprisingly rapidly, this is an undoubted shortcoming. In principle, not all the $\omega_{i,j}$ s in (1.1) need be large and we do not wish to employ different quadrature rules for different $\omega_{i,j}$ s, something that unduly complicates things.

Fortunately, a major lesson of recent advances in numerical quadrature of highly oscillatory integrals is that the main role of asymptotic formulæ like (2.7) is as a gateway to Filon-type [11] and Levin-type [16] techniques. An important advantage of these methods is that they segue seamlessly into classical quadrature for small ω , hence are uniformly effective throughout the entire range of frequencies. They are also typically significantly more accurate that the asymptotic method (2.8).

3. A FILON-TYPE METHOD

An alternative to the asymptotic method (2.8) is a Filon-type method. Thus, let $\nu \geq 2$, nodes $-1 = c_1 < c_2 < \ldots < c_{\nu} = 1$ and multiplicities $m_1, m_2, \ldots, m_{\nu} \in \mathbb{N}$. We interpolate the function f in a Hermite sense at the nodes c by a polynomial p of degree $\sum_{i=1}^{\nu} m_i - 1$,

$$p^{(j)}(c_k) = f^{(j)}(c_k), j = 0, \dots, m_k - 1, k = 1, 2, \dots, \nu.$$
 (3.1)

The Filon-type method for the highly oscillatory integral (2.1) is defined as

$$\mathbf{F}[f] = \mathbf{I}[p] = \int_{-1}^{1} p(x) e^{\tau \sin \omega (\alpha x + \beta)} dx.$$
 (3.2)

Theorem 3.1. Let $\alpha \neq 0$ and $s = \min\{m_1, m_{\nu}\}$. Then for every $f \in C^{\infty}[-1, 1]$

$$F[f] - I[f] \sim I_0(\tau)E[f] + \mathcal{O}(\omega^{-s-1}), \qquad \omega \gg 1,$$
 (3.3)

where $E[f] = \int_{-1}^{1} [p(x) - f(x)] dx$.

Proof. We use the method of proof from [11]. Since both F and I are linear operators, F[f] - I[f] = I[p-f] and the theorem follows at once from letting p-f in (2.8) and noting that the interpolation conditions (3.1) annihilate asymptotic terms for $m=0,1,\ldots,\lfloor (s-1)/2\rfloor$ in the first sum in (2.7) and $m=0,1,\ldots,\lfloor s/2\rfloor-1$ in the second.

Note that the internal nodes $c_2, \ldots, c_{\nu-1}$ have no influence upon the asymptotic order of the error. However, they have three important functions. Firstly, good choice of such points minimizes the non-oscillatory quadrature error E[f], one of the two components of the quadrature error in (3.3). Secondly, intuitively speaking,

the method (3.2) is nothing but the asymptotic quadrature A_s , applied to the interpolation error p-f rather than to the original function f. Thus, the smaller we make the interpolation error, the better. Thirdly, unlike (2.8), the Filon-type method is relevant throughout the range of frequencies $\omega \in \mathbb{R}$. In particular, when $|\omega|$ is small then $F[f] = E[f] + \mathcal{O}(\omega)$, the reason being that $I[f] = \int_{-1}^{1} f(x) \, dx + \mathcal{O}(\omega)$ and $F[f] = I[p] = \int_{-1}^{1} p(x) \, dx + \mathcal{O}(\omega) = Q[f] + \mathcal{O}(\omega)$, where Q[f] (upon which we will dwell further in Sect. 3.2) is the classical Birkhoff-Hermite quadrature induced by our nodes and multiplicities. Thus, rendering |E[f]| small is vital also in this regime.

3.1. Implementation of the Filon-type method

The implementation of (3.2) is based on the premise that we can integrate (2.1) exactly once f is a polynomial. Thus, let

$$p(x) = \sum_{r=0}^{q} p_r x^r$$
, where $q = \sum_{i=1}^{\nu} m_i - 1$.

Then

$$F[f] = \sum_{r=0}^{q} p_r \int_{-1}^{1} x^r e^{\tau \sin \omega (\alpha x + \beta)} dx = \sum_{r=0}^{q} p_r \mu_r(\omega).$$
 (3.4)

The moments μ_r can be calculated directly from the asymptotic expansion (2.7) since the latter terminates in that case,

$$\begin{split} \mu_r(\omega) &= \frac{1 + (-1)^r}{r+1} I_0(\tau) \\ &+ \sum_{m=0}^{\lfloor r/2 \rfloor} \frac{(-1)^m}{(\alpha \omega)^{2m+1}} \frac{r!}{(r-2m)!} \left\{ \left[\Theta_m^{[2]}(\omega(\alpha+\beta)) - \Theta_m^{[1]}(\omega(\alpha+\beta)) \right] \right. \\ &+ (-1)^r \left[\Theta_m^{[2]}(\omega(\alpha-\beta)) + \Theta_m^{[1]}(\omega(\alpha-\beta)) \right] \right\} \\ &+ \sum_{m=0}^{\lfloor (r-1)/2 \rfloor} \frac{(-1)^m}{(\alpha \omega)^{2m+2}} \frac{r!}{(r-2m-1)!} \left\{ \left[\Phi_m^{[2]}(\omega(\alpha+\beta)) + \Phi_m^{[1]}(\omega(\alpha+\beta)) \right] \right. \\ &+ (-1)^r \left[\Phi_m^{[2]}(\omega(\alpha-\beta)) - \Phi_m^{[1]}(\omega(\alpha-\beta)) \right] \right\}, \qquad r \in \mathbb{Z}_+. \end{split}$$

Note that (3.4) is not a practical means to calculate F[f]. Like in the case of non-oscillatory quadrature, it is advantageous to express p in terms of cardinal polynomials,

$$p(x) = \sum_{k=1}^{\nu} \sum_{j=0}^{m_k - 1} \ell_{k,j}(x) f^{(j)}(c_k),$$

where each $\ell_{k,j}$ is a polynomial of degree q such that

$$\ell_{k,j}^{(i)}(c_n) = \begin{cases} 1, & k = n, \ i = j, \\ 0, & \text{otherwise} \end{cases}$$
 (3.5)

for $i = 0, 1, ..., m_n - 1, j = 0, 1, ..., m_k - 1, k, n = 1, 2, ..., \nu$. Letting

$$b_{k,j} = \mathbf{I}[\ell_{k,j}] = \int_{-1}^{1} \ell_{k,j}(x) e^{\tau \sin(\omega(\alpha x + \beta))} dx, \quad j = 0, 1, \dots, m_k - 1, \quad k = 1, 2, \dots, \nu$$

(which we can do once-for-all in terms of the moments μ_r) we obtain

$$\mathbf{F}[f] = \sum_{k=1}^{\nu} \sum_{j=0}^{m_k - 1} b_{k,j} f^{(j)}(c_k), \tag{3.6}$$

a form reminiscent of classical quadrature [4].

To illustrate our construction by few simple examples, let us assume (mostly to render notation more transparent) that $\alpha = 1$ and $\beta = 0$, whereby

$$\begin{split} \mu_0(\omega) &= 2I_0(\tau) + \frac{2}{\omega}\Theta_0^{[2]}(\omega), \\ \mu_1(\omega) &= -\frac{2}{\omega}\Theta_0^{[1]}(\omega) + \frac{2}{\omega^2}\Phi_0^{[1]}(\omega), \\ \mu_2(\omega) &= \frac{2}{3}I_0(\tau) + \frac{2}{\omega}\Theta_0^{[2]}(\omega) + \frac{4}{\omega^2}\Phi_0^{[2]}(\omega) - \frac{4}{\omega^3}\Theta_1^{[2]}(\omega), \\ \mu_3(\omega) &= -\frac{2}{\omega}\Theta_0^{[1]}(\omega) + \frac{6}{\omega^2}\Phi_0^{[1]}(\omega) + \frac{12}{\omega^3}\Theta_1^{[1]}(\omega) - \frac{12}{\omega^4}\Phi_1^{[1]}(\omega), \\ \mu_4(\omega) &= \frac{2}{5}I_0(\tau) + \frac{2}{\omega}\Theta_0^{[2]}(\omega) + \frac{8}{\omega^2}\Phi_0^{[2]}(\omega) - \frac{24}{\omega^3}\Theta_1^{[2]}(\omega) - \frac{48}{\omega^4}\Phi_1^{[2]}(\omega) + \frac{48}{\omega^5}\Theta_2^{[2]}(\omega). \end{split}$$

(Note that we have suppressed the dependence of $\Phi_m^{[i]}$ and $\Theta_m^{[i]}$ on τ .)

We commence from $\nu=2$, $\boldsymbol{c}=[-1,1]$ and $\boldsymbol{m}=[1,1]$, whereby $p(x)=\frac{1}{2}(1-x)f(-1)+\frac{1}{2}(1+x)f(1)$. This results in the method

$$\mathbf{F}[f] = \frac{1}{2} [\mu_0(\omega) - \mu_1(\omega)] f(-1) + \frac{1}{2} [\mu_0(\omega) + \mu_1(\omega)] f(1).$$

Next, we consider c = [-1, 1] and m = [2, 2], whereby

$$p(x) = \frac{1}{4}(1+x)(2+x-x^2)f(1) + \frac{1}{4}(1-x)(2-x-x^2)f(-1) - \frac{1}{4}(1-x)(1+x)^2f'(1) + \frac{1}{4}(1-x)^2(1+x)f'(-1)$$

and

$$F[f] = \frac{1}{4}(2\mu_0 + 3\mu_1 - \mu_3)f(1) + \frac{1}{4}(2\mu_0 - 3\mu_1 + \mu_3)f(-1) - \frac{1}{4}(\mu_0 + \mu_1 - \mu_2 - \mu_3)f'(1) + \frac{1}{4}(\mu_0 - \mu_1 - \mu_2 + \mu_3)f'(-1).$$

As our final example, we let $\nu = 3$, c = [-1, 0, 1] and m = [2, 1, 2]. We now have

$$p(x) = \frac{1}{4}x(1+x)^2(3-2x)f(1) + (1-x^2)^2f(0) - \frac{1}{4}x(1-x)^2(3+2x)f(-1) - x(1-x)(1+x)^2f'(1) - x(1-x)^2(1+x)f'(-1),$$

therefore

$$F[f] = \frac{1}{4}(3\mu_1 + 4\mu_2 - \mu_3 - 2\mu_4)f(1) + (\mu_0 - 2\mu_2 + \mu_4)f(0) + \frac{1}{4}(-3\mu_1 + 4\mu_2 + \mu_3 - 2\mu_4)f(-1) + \frac{1}{4}(-\mu_1 - \mu_2 + \mu_3 + \mu_4)f'(1) + \frac{1}{4}(-\mu_1 + \mu_2 + \mu_3 - \mu_4)f'(-1).$$

3.2. Hermite-Birkhoff quadrature

Wishing to minimise the non-oscillatory error E[f], we have the freedom of choosing nodes and weights, subject to $c_1 = -1$, $c_{\nu} = 1$ and $s = \min\{m_1, m_{\nu}\}$, a procedure that has been already considered in [12]. Let

$$\tilde{b}_{k,j} = \int_{-1}^{1} \ell_{k,j}(x) \, \mathrm{d}x, \qquad j = 0, 1, \dots, m_k - 1, \quad k = 1, 2, \dots, \nu,$$

where the $\ell_{k,j}$ s were defined in (3.5), and

$$Q[f] = \sum_{k=1}^{\nu} \sum_{j=0}^{m_k - 1} \tilde{b}_{k,j} f^{(j)}(c_k), \tag{3.7}$$

we have $E[f] = Q[f] - \int_{-1}^{1} f(x) dx$. Therefore Q is a Hermite-Birkhoff quadrature [15] for the computation of a non-oscillatory integral.

Theorem 3.2. Let $m_1 = m_{\nu} = s$ and $m_2 = m_3 = \ldots = m_{\nu-1} \equiv 1$. The quadrature (3.7) is of maximal order $2\nu + 2s - 4$ (in other words, is exact for all polynomials of degree $2\nu + 2s - 5$ when $c_2, c_3, \ldots, c_{\nu-1}$ are the zeros of the Jacobi polynomial $P_{\nu-2}^{(s,s)}$).

Proof. A straightforward generalisation of the familiar proof on the order of Gauss-Christoffel quadrature [4]. Let

$$\mathbb{P}_{\nu+2s-2}[x] \ni u(x) = (1-x^2)^s P_{\nu-2}^{(s,s)}(x),$$

where $\mathbb{P}_n[x]$ is the set of *n*th-degree polynomials. Given any $w \in \mathbb{P}_{2\nu+2s-5}[x]$, it follows by the Euclidean algorithm that there exist $p \in \mathbb{P}_{\nu-3}[x]$ and $q \in \mathbb{P}_{\nu+2s-3}[x]$ such that w = pu + q. Recalling that $P_{\nu-2}^{(s,s)}$ is orthogonal in (-1,1) with respect to the weight function $(1-x^2)^s$ [1], we have

$$\int_{-1}^{1} p(x)u(x) dx = \int_{-1}^{1} p(x)P_{\nu-2}^{(s,s)}(x)(1-x^{2})^{s} dx = 0,$$

while Q[pu] = 0 because $u(c_k) = 0, k = 1, 2, ..., \nu$ and $u^{(j)}(\pm 1) = 0, j = 0, 1, ..., s - 1$. Therefore

$$\boldsymbol{E}[w] = \boldsymbol{Q}[w] - \int_{-1}^{1} w(x) dx = \boldsymbol{Q}[q] - \int_{-1}^{1} q(x) dx.$$

The right-hand side vanishes because the weights are interpolatory. This is standard argument in classical quadrature and follows in a Hermite-Birkhoff setting by counting $\nu + 2s$ degrees of freedom and observing that the underlying linear system (ensuring that Q is at least of order $\nu + 2s$) is nonsingular, being a limiting case of Lagrangian interpolation with $\nu + 2s$ nodes. We deduce that E[w] = 0 for every $w \in \mathbb{P}_{2\nu+2s-5}[x]$, hence order $2\nu + 2s - 4$.

It remains to prove that no other choice of internal nodes $c_2, c_3, \ldots, c_{\nu-1}$ may increase the order. To this end it is sufficient to single out one polynomial $v \in \mathbb{P}_{2\nu+2s-4}[x]$ such that $\boldsymbol{E}[v] \neq 0$ for any choice of internal nodes. We thus choose

$$v(x) = (1 - x^2)^s \prod_{k=2}^{\nu-1} (x - c_k)^2,$$

where $c_2, c_3, \ldots, c_{\nu-1} \in (-1, 1)$ are arbitrary. Trivially, $\int_{-1}^1 v(x) dx > 0$. On the other hand, $v(c_k) = 0$, $k = 1, 2, \ldots, \nu$ and $v^{(j)}(\pm 1) = 0$, $j = 0, 1, \ldots, s-1$, imply that $\mathbf{Q}[v] = 0$. It thus follows that $\mathbf{E}[v] < 0$ and the maximal order is indeed $2(\nu + s - 2)$.

TABLE 3. Absolute errors in approximating $\int_{-1}^{1} e^x dx$ by $\mathbf{Q}[e^x]$ with $s = m_1 = m_{\nu} = 2$ (top), $s = m_1 = m_{\nu} = 3$ (bottom), $m_2 = m_3 = \ldots = m_{\nu-1} = 1$ and $\nu = 2, \ldots, 7$.

| ν | 2 | 3 | 4 | 5 | 6 | 7 |
|-------|--------------|--------------|--------------|--------------|--------------|--------------|
| s=2 | 4.77_{-02} | 2.21_{-04} | 7.42_{-07} | 1.74_{-09} | 2.93_{-12} | 3.71_{-15} |
| s = 3 | 1.34_{-03} | 2.61_{-06} | 4.65_{-09} | 6.61_{-12} | 7.43_{-15} | 6.77_{-18} |

To flesh out numbers, herewith few explicit quadratures (3.7):

$$\begin{split} \nu &= 2, \ s = 2 \colon \qquad \mathbf{Q}[f] = f(1) + f(-1) - \frac{1}{3}[f'(1) - f'(-1)], \\ \nu &= 2, \ s = 3 \colon \qquad \mathbf{Q}[f] = f(1) + f(-1) - \frac{2}{3}[f'(1) - f'(-1)] + \frac{1}{15}[f''(1) + f''(-1)], \\ \nu &= 3, \ s = 2 \colon \qquad \mathbf{Q}[f] = \frac{1}{15}[7f(1) + 16f(0) + 7f(-1)] - \frac{1}{15}[f'(1) - f'(-1)], \\ \nu &= 4, \ s = 2 \colon \qquad \mathbf{Q}[f] = \frac{1}{135}[37f(1) + 98f(\frac{\sqrt{7}}{7}) + 98f(-\frac{\sqrt{7}}{7}) + 37f(-1)] \\ &\qquad - \frac{1}{45}[f'(1) - f'(-1)], \\ \nu &= 3, \ s = 3 \colon \qquad \mathbf{Q}[f] = \frac{1}{35}[19f(1) + 32f(0) + 19f(-1)] - \frac{4}{35}[f'(1) - f'(-1)] \\ &\qquad + \frac{1}{105}[f''(1) + f''(-1)], \\ \nu &= 4, \ s = 3 \colon \qquad \mathbf{Q}[f] = \frac{1}{1120}[391f(1) + 729f(\frac{1}{3}) + 729f(-\frac{1}{3}) + 391f(-1)] \\ &\qquad - \frac{13}{280}[f'(1) - f'(-1)] + \frac{1}{420}[f''(1) + f''(-1)]. \end{split}$$

Because of the use of zeros of Jacobi polynomials as internal nodes, according to Theorem 3.2 the order in each case is $2(\nu + s - 2)$.

Of course, there is nothing to prevent us from using higher multiplicities with internal nodes, except that we might lose the attractive feature of Theorem 3.2, reminiscent of the Gauss-Christoffel quadrature, namely that maximal order exceeds by ν the number of degrees of freedom. Thus, for example, choosing $\nu = 3$, c = [-1, 0, 1] and $m_k \equiv 2$, the coefficient of f'(0) is nil and we recover the sixth-order formula with $\nu = 3$, s = 1, above. On the other hand, once we let m = [2, 3, 2], we obtain

$$Q[f] = \frac{1}{35} [11f(1) + 48f(0) + 11f(-1)] - \frac{1}{35} [f'(1) - f'(-1)] + \frac{8}{105} f''(0),$$

of order 8

In Table 3 we display errors E[f] committed by Hermite-Birkhoff methods consistent with the conditions of Theorem 3.2, with s=2 and increasing values of ν , applied to the function $f(x)=\mathrm{e}^x$. The decrease in error is consistent with Theorem 3.2.

3.3. Numerical examples for Filon-type methods

According to (3.3), the error of Filon-type methods has two components. The asymptotic component decays with increasing ω but $I_0(\tau)E[f]$ is independent of ω . Thus, unlike in the case of Filon-type methods for 'classical' highly oscillatory integrals [11] and in variance with the asymptotic method A[f], the error does not tend to zero for $\omega \to \infty$. This is demonstrated in Figure 6, where we display the absolute error F[f] - I[f] for the Filon-type

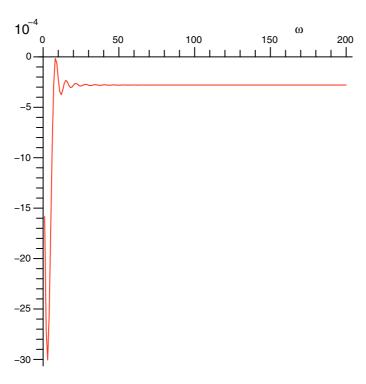


FIGURE 6. The error $F[e^x] - I[e^x]$ for c = [-1, 0, 1], m = [2, 1, 2] and $\omega \in [0, 200]$.

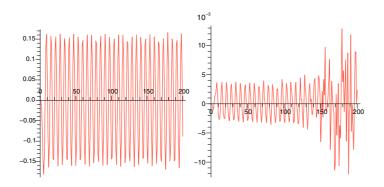


FIGURE 7. Scaled errors $\omega^3 | \boldsymbol{F}[\mathbf{e}^x] - \boldsymbol{I}[\mathbf{e}^x] - I_0(1) E[\mathbf{e}^x] | \text{ for } \boldsymbol{c} = [-1, 0, 1], \ \boldsymbol{m} = [2, 1, 2] \text{ (left)}, \ \boldsymbol{c} = [-1, -\frac{\sqrt{7}}{7}, \frac{\sqrt{7}}{7}, 1], \ \boldsymbol{m} = [2, 1, 1, 2] \text{ (right) and } \omega \in [0, 200].$

method with $\boldsymbol{c}=[-1,0,1], \, \boldsymbol{m}=[2,1,2]$ and the function $f(x)=\mathrm{e}^x$. For $\omega\gg 1$ the error asymptotes to $\approx -2.79_{04}=I_0(1)E[\mathrm{e}^x]$ (cf. Tab. 3, $\nu=3$, for $E[\mathrm{e}^x]$).

In Figure 7 we display the asymptotic error component $F[f] - I[f] - I_0(\tau)E[f]$, scaled by ω^3 , for two different Filon-type methods of an asymptotic order two. In both cases, consistently with Theorem 3.2, the scaled error asymptotes to a constant.

It is instructive to compare absolute errors at different values of ω for asymptotic and Filon-type methods. In an important aspect, this comparison is heavily weighed against Filon-type methods, because the asymptotic method (2.8) assumes that $\int_{-1}^{1} f(x) dx$ is calculated exactly: in practice we need to replace the integral

Table 4. Absolute errors $|F[e^x] - I[e^x]|$ for different Filon-type methods.

by quadrature. Nonetheless, and even bearing in mind that E[f] makes up the major share of error in (3.2), Filon-type methods are evident superior.

The error for asymptotic methods is displayed in Table 2 and, predictably, it starts unacceptably high but becomes increasingly small, tending to zero for $\omega \gg 1$. Not so for Filon-type methods, exhibited in Table 4. The uniform error of (3.2) is considerably smaller, because the performance for small and moderate ω s is considerably better. On the other hand, the error for $\omega \gg 1$ does not tend to zero, as we have already repeatedly observed. Overall, it is clear that Filon-type methods significantly decrease the error at the cost of few extra function evaluations, even when the integral in $A_s[f]$ is computed exactly.

We note in passing that the fixed error component E[f] assumes significantly smaller importance in the setting of ordinary differential equations and the solution of the integral (1.2). In that case E[f] is scaled by h^q , where $h = t_{n+1} - t_n$ is the length of the integration interval and q is the order of the Hermite-Birkhoff quadrature. In this setting Filon-type methods are likely to outperform asymptotic methods by a large margin, since the latter are largely insensitive to the length of integration integral.

4. Numerical examples

We bring the equation (1.2) into a form appropriate for the application of Filon-type methods in the interval [-1, 1],

$$\mathbf{y}(t_{n+1}) = e^{hA} \mathbf{y}(t_n) + \frac{h}{2} \int_{-1}^{1} e^{\frac{1}{2}h(1-x)A} E(h(n+\frac{1}{2}+\frac{1}{2}x)) \mathbf{g}(h(n+\frac{1}{2}+\frac{1}{2}x)) dx.$$
(4.1)

Our time-stepping routine is obtained by replacing integrals with appropriate Filon-type methods.

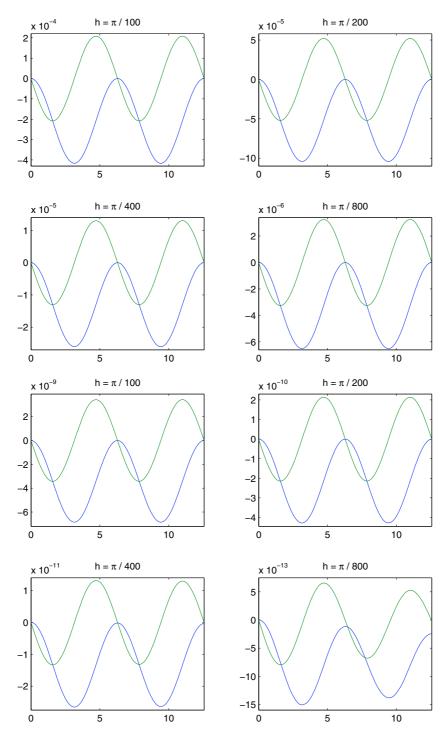
In the specific context of equation (1.4), the time-stepping formula (4.1), combined with a Filon-type solver, becomes

$$y_{n+1,1} = y_{n,1}\cos h + y_{n,2}\sin h + h\mathbf{F}[\sin(\frac{1}{2}h(1-x))],$$

$$y_{n+1,2} = -y_{n,1}\sin h + y_{n,2}\cos h + h\mathbf{F}[\cos(\frac{1}{2}h(1-x))],$$

where the Filon-type methods are applied with $\alpha = \frac{1}{2}, \beta = n + \frac{1}{2}$ and ω replaced by $h\omega$. The pointwise error for two Filon-type methods is displayed in Figure 8 for step-sizes $h \in \{\frac{\pi}{100}, \frac{\pi}{200}, \frac{\pi}{400}, \frac{\pi}{800}\}$. (We have used there the more precise Filon-type method with step-size $h = \frac{\pi}{1600}$ as our 'true' solution.) The first Filon-type method uses only function values at the endpoints, the second uses both function values and derivatives there. (We did not use any internal points but note in passing that their incorporation would have further reduced the error.) A comparison with Table 1 is striking: at the cost of just 400 steps with the plainvanilla Filon-type method (requiring just one new function evaluation per step!) we produce better accuracy than ode45 with 240645 steps.

Note that the errors in Figure 8 appear to be the same periodic function, scaled by suitable powers of h(except for the second method with $h = \frac{\pi}{800}$, but this is likely to be a machine-precision artifact). So is the exact solution (cf. bottom of Fig. 1) but these two functions are different. The reason for periodicity can tell



 $\ensuremath{\mathsf{Figure}}$ 8. Pointwise absolute error for two Filon-type methods.

us something about the properties of our method, hence it bears some elaboration. Recall that for large ω the major source for the error is the classical quadrature error (scaled by a suitable Bessel function). Let $e_n = y_n - y(t_n)$ and consider just the error originating in classical quadrature. It readily follows that

$$\boldsymbol{e}_{n+1} = e^{hA}\boldsymbol{e}_n + h\boldsymbol{q},\tag{4.2}$$

where the vector \mathbf{q} contains the contribution of classical quadrature error for the different components. For example, in the present case, for 'plain-vanilla' Filon we have

$$q = I_0(1) \begin{bmatrix} h \sin h - 2 + 2 \cos h \\ h + h \cos h - 2 \sin h \end{bmatrix} = \begin{bmatrix} \mathcal{O}(h^4) \\ \mathcal{O}(h^3) \end{bmatrix}.$$

Bearing in mind that $e_0 = 0$, the solution of (4.2) is

$$\boldsymbol{e}_n = h(I - e^{hA})^{-1}(I - e^{nhA})\boldsymbol{q},$$

which in the present case becomes

$$\boldsymbol{e}_n = \frac{h}{2(1-\cos h)} \left[\begin{array}{ccc} 1-\cos h & \sin h \\ -\sin h & 1-\cos h \end{array} \right] \left[\begin{array}{ccc} 1-\cos nh & -\sin nh \\ \sin nh & 1-\cos nh \end{array} \right] \boldsymbol{q}.$$

Periodicity is clear, as is the fact that in Figure 6 the error is a linear combination of just two harmonics, $\sin nh$ and $\cos nh$.

Be it as may, it is crystal clear that even the most elementary Filon-type methods enjoy tremendous advantage in comparison to state-of-the-art general ODE software like ode45 when applied with high frequencies. Another important advantage of Filon-type methods, which is not apparent from our comparison, is that both the error and the computational effort are roughly uniform in frequency, while classical ODE solvers deteriorate with increasing frequency. Note that we have implemented Filon-type methods in the most straightforward manner, with constant step size and without any error control (cf. [10] for error control for Filon-type methods). It is highly likely that more sophisticated implementation would have resulted in even more striking outcome.

It is equally clear that our analysis is far of complete. Modern methods for highly oscillatory differential equations set themselves ambitious goals: not just time-stepping with a step size that does not decrease unduly in the presence of high frequencies but also correct long-term asymptotic behaviour, good control of global error and the conservation of important qualitative and geometric attributes of the exact solution [2,8]. This is in keeping with the character of this paper as an initial foray into the subject matter of using modern computational techniques for the highly oscillatory problems occurring in electronic engineering.

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References

- [1] M. Abramowitz and I.A. Stegun, *Handbook of Mathematical Functions*. National Bureau of Standards, Washington, DC, (1964).
- [2] D. Cohen, T. Jahnke, K. Lorenz and C. Lubich, Numerical integrators for highly oscillatory Hamiltonian systems: a review, in *Analysis, Modeling and Simulation of Multiscale Problems*, A. Mielke Ed., Springer-Verlag (2006) 553–576.
- [3] E. Dautbegovic, M. Condon and C. Brennan, An efficient nonlinear circuit simulation technique. *IEEE Trans. Microwave Theory Tech.* **53** (2005) 548–555.
- [4] P.J. Davis and P. Rabinowitz, Methods of Numerical Integration. Second Edition, Academic Press, Orlando, USA (1984).
- [5] V. Grimm and M. Hochbruck, Error analysis of exponential integrators for oscillatory second-order differential equations. J. Phys. A: Math. Gen. 39 (2006) 5495–5507.

- [6] S. Haykin, Communications Systems. Fourth Edition, John Wiley, New York, USA (2001).
- [7] D. Huybrechs and S. Vandewalle, On the evaluation of highly oscillatory integrals by analytic continuation. SIAM J. Numer. Anal. 44 (2006) 1026–1048.
- [8] A. Iserles, On the global error of discretization methods for highly-oscillatory ordinary differential equations. BIT 42 (2002a) 561–599.
- [9] A. Iserles, Think globally, act locally: solving highly-oscillatory ordinary differential equations. Appl. Num. Anal. 43 (2002b) 145–160.
- [10] A. Iserles and S.P. Nørsett, On quadrature methods for highly oscillatory integrals and their implementation. BIT 44 (2004) 755–772.
- [11] A. Iserles and S.P. Nørsett, Efficient quadrature of highly oscillatory integrals using derivatives. Proc. Royal Soc. A 461 (2005) 1383–1399.
- [12] A. Iserles and S.P. Nørsett, From high oscillation to rapid approximation I: Modified Fourier expansions. IMA J. Num. Anal. 28 (2008) 862–887.
- [13] M.C. Jeruchim, P. Balaban and K.S. Shanmugan, Simulation of Communication Systems, Modeling, Methodology and Techniques. Second Edition, Kluwer Academic/Plenum Publishers, New York, USA (2000).
- [14] M. Khanamirian, Quadrature methods for systems of highly oscillatory ODEs. Part I. BIT 48 (2008) 743-761.
- [15] C.A. Micchelli and T.J. Rivlin, Quadrature formulæ and Hermite-Birkhoff interpolation. Adv. Maths 11 (1973) 93–112.
- [16] S. Olver, Moment-free numerical integration of highly oscillatory functions. IMA J. Num. Anal. 26 (2006) 213–227.
- [17] R. Pulch, Multi-time scale differential equations for simulating frequency modulated signals. Appl. Numer. Math. 53 (2005) 421–436.
- [18] J. Roychowdhury, Analysing circuits with widely separated time scales using numerical PDE methods. *IEEE Trans. Circuits Sys. I, Fund. Theory Appl.* **48** (2001) 578–594.
- [19] C.J. Weisman, The Essential Guide to RF and Wireless. Second Edition, Prentice-Hall, Englewood Cliffs, USA (2002).
- [20] R. Wong, Asymptotic Approximations of Integrals. SIAM, Philadelphia (2001).